# Impact of Relaxation on the Performance of GeSe True Random Number Generator Based on Ovonic Threshold Switching 

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#### Abstract

Volatile Ovonic threshold switching (OTS) are promising not only as the selector in crossbar resistive switching memory arrays, but also as true random number generators (TRNG) by utilizing its probabilistic switching characteristics. However, investigation on the reliability of OTS-based TRNG is still lacking, which hinders its practical application. Previously, we found that switching probability is dependent on the pulse amplitude and width. In this work, we report that relaxation which happens during the time interval between pulses can also cause switching probability drift. Optimizing the bit-generation waveform and modulating the pulse conditions could provide a practical solution, in addition to the impact of external bias and temperature. This work provides useful guidance for the practical design and operation of OTS-based TRNGs.


Index Terms-OTS, GeSe, random number generation, relaxation, threshold voltage.

## I. Introduction

RANDOM number generators (RNGs) are playing pivotal roles in statistical sampling [1], [2], computer simulation [3], cryptography [4], [5], etc. Conventionally, random bit streams are generated using software-based pseudo-RNGs, and inevitably contain repetition and correlation [6]. In contrast, a true random number generator (TRNG) uses physical

[^0]phenomena as sources of randomness and cannot be replicated or predicted externally. The existing CMOS-based TRNGs are normally based on white noise [7], random telegraph noise (RTN) [8], [9], time-dependent dielectric breakdown [10], or some other noise sources [11]-[16]. However, in those TRNG implementations, complex signal post-processing systems are required, leading to increasing scaling difficulty.

Earlier works have reported that the Ovonic threshold switching (OTS) chalcogenide materials, a promising material for selector applications in emerging memory arrays [17]-[20], can also work as a randomness source by exploiting its randomly distributed time-to-switch-on/off [21]-[23]. The volatile nature of OTS offers an important advantage, as it avoids the reset operation in TRNGs based on non-volatile memory devices for erasing the previous generated bit, thus simplifying the circuitry, and improving the throughput. Our preliminary exploration of OTS-based TRNG shows satisfying randomness in 10,000 bits [21] and its promising potential in the application of stochastic computing [23]. However, further reliability investigation under variable operation conditions is still needed for the practical TRNG application.

In this work, we further look into the switching probability drift in OTS-based TRNGs using statistical measurements, and attributed it to the relaxation phenomenon in chalcogenide after switching, which leads to threshold voltage $\left(\mathrm{V}_{\mathrm{th}}\right)$ drift. Such relaxation can be accelerated by using external bias or higher temperature, however, optimizing the bit-generation waveform and modulating the pulse conditions is a more practical solution for TRNG circuitry. This work provides useful guidance for the practical design and operation of OTS-TRNGs.

## II. Device and Characterization

Amorphous $\mathrm{Ge}_{\mathrm{x}} \mathrm{Se}_{1-\mathrm{x}}$ films were sandwiched between two TiN electrodes to form a $\mathrm{TiN} / \mathrm{Ge}_{\mathrm{x}} \mathrm{Se}_{1-\mathrm{x}} / \mathrm{TiN}$ structure. The OTS devices were integrated in a 300 nm process flow and device size is defined by the TiN bottom electrode [24]. The $\mathrm{Ge}_{\mathrm{x}} \mathrm{Se}_{1-\mathrm{x}}$ chalcogenide films were deposited and passivated by a low-temperature BEOL process scheme, as shown in Fig. 1a. The device size is 65 nm and the $\mathrm{Ge}_{\mathrm{x}} \mathrm{Se}_{1-\mathrm{x}}$ thickness is 10 nm . The fast I-V characterization was done with a Keysight B1500A semiconductor analyzer with embedded B1530A Waveform Generator/Fast Measurement Unit (WGFMU). Fig. 1b schematizes the random generation


Fig. 1. (a) TEM of a $\mathrm{Ge}_{\mathrm{x}} \mathrm{Se}_{1-\mathrm{x}}$ OTS selector. (b) Schematic of the random generation procedure, including a waveform generator, an OTS selector and a comparator to provide digital " 1 " and " 0 ". Current is measured at the end of each pulse. Digital bits " 1 " or " 0 " are generated by a comparator, depending whether OTS has been switched on or not.


Fig. 2. Demonstration of the current measured in 1000 pulses $\left(\mathrm{V}_{\text {pulse }}=\right.$ 2.7 V , $\mathrm{t}_{\text {pulse }}=1 \mu \mathrm{~s}$ ), with interval time of (a) $1 \mu \mathrm{~s}$ and (b) 1 s respectively. The switching probability drops significantly with longer interval time.
procedure in our previous work [21], [23]: A sequence of pulses generated by a waveform generator is applied onto the top electrode of OTS, while the current is measured from the bottom electrode at a certain point of each pulse. Digital bits " 1 " or " 0 " are generated by a comparator, depending whether OTS has been switched on at this measurement point.

## III. Results and Discussions

Our previous works revealed that the stochastic switching of OTS originates from the Weibull-distribution of time-to-switch-on (t-on) [21], [23], [25]. Particularly, [23] shows that switching probability can be tuned by the pulse amplitude or pulse width. For a practical TRNG, it is also desirable that the TRNG works stably under different pulse intervals, i.e. the switching probability does not drift with different interval time between pulses. Unfortunately, using the previously proposed setup [21], [23], the switching probability drops significantly if the pulse interval increases from $1 \mu \mathrm{~s}$ to 1 s (Fig.2), with a fixed pulse amplitude of 2.7 V and pulse width of $1 \mu \mathrm{~s}$.

The impact of interval time is further analyzed using a train of triangular pulses with the same amplitude ( 5 V ) and rise/fall time ( $1 \mu \mathrm{~s}$ ) but increasing interval times (Fig. 3a). The benefit of using triangular pulses is to precisely monitor the voltage as well as the time in OTS switching [17], [25]. A higher amplitude of 5.5 V is applied to first-fire a fresh device and 100 triangular pulses with interval of $1 \mu \mathrm{~s}$ are then applied to stabilize the device. Following that, the triangular pulse train in Fig. 3a starts. The $1^{\text {st }} \mathrm{V}_{\text {th }}$ is around 2.9 V , while the $2^{\text {nd }}$ Vth after an interval of $1 \mu$ s becomes much lower (2.5 V). However, the following $\mathrm{V}_{\text {th }} \mathrm{s}$ gradually "recovers" after longer interval time, and could even surpass the $1^{\text {st }} \mathrm{V}_{\text {th }}$ if the interval reaches 1000 s (Fig. 3b). Such $\mathrm{V}_{\text {th }}$ recovery is statistically verified using a designed waveform with two


Fig. 3. (a) A train of triangular pulses with increasing interval times to analyse the impact of interval time on $\mathrm{V}_{\text {th }}$. (b) $\mathrm{V}_{\text {th }}$ increases with longer relaxation time. (c) Statistical distribution of $\mathrm{V}_{\text {th }}$ with different interval time measured using the (inset) sandwiched waveform.


Fig. 4. Relaxation under different temperature and bias conditions, under sandwiched triangular pulses in Fig. 3c. (a) The median $\left(\mathrm{V}_{\mathrm{th} 2}-\mathrm{V}_{\mathrm{th} 1}\right) / \mathrm{V}_{\text {th1 }}$ at different temperatures with $\mathrm{V}_{\text {relax }}$ fixed at OV and relaxation time of 1 $\mu \mathrm{s}$. (b) The median $\left(\mathrm{V}_{\text {th2 }}-\mathrm{V}_{\text {th1 }}\right) / \mathrm{V}_{\text {th1 }}$ at different $\mathrm{V}_{\text {relax }}$ at $25^{\circ} \mathrm{C}$ with relaxation time of $1 \mu \mathrm{~s}, 1 \mathrm{~ms}$ and 1 s .
triangular pulses sandwiched with an interval period (inset in Fig. 3c). For each interval time, the sandwiched waveform is repeated for 100 times, with a fixed gap period of 10 s in between. Obviously, for each interval time, $\mathrm{V}_{\text {th }}$ follows normal distribution, which shifts rightward with longer interval times, (Fig. 3c).

Such $\mathrm{V}_{\text {th }}$ recovery phenomenon could explain the change in switching probability at longer intervals proposed in Fig. 2: (i) the longer interval time between each square pulse leads to a higher $\mathrm{V}_{\text {th }}$; (ii) moreover, a failure-to-switch-on by one pulse effectively prolongs the relaxation period and further increases the $\mathrm{V}_{\mathrm{th}}$, making it more difficult to switch-on by the next pulse. Therefore, the relaxation issue could be the main reason for the probability drift of OTS-based TRNG when interval time varies.

The $\mathrm{V}_{\text {th }}$ recovery of OTS is associated with the delocalization of defects [26]. After the switch-off, some defects remain delocalized, leading to a reduced $\mathrm{V}_{\text {th }}$ in subsequent switching. During relaxation, the remaining delocalized defects gradually return to the localized state, and thus a higher voltage is needed to delocalize those defects, leading to a higher $\mathrm{V}_{\mathrm{th}}$ [17].

## Optimized bit generation waveform



Fig. 5. (a) The optimized 2-stage "square-triangular" waveform to remove the cumulative effect of failure-to-switch-on. Current is measured at the end of the square pulse. (b) Comparison of the switching probability measured with various interval time (○) using square pulse of 2.7 V , ( $\square$ ) modulated square pulse of $2.7 \mathrm{~V}, 2.8 \mathrm{~V}$ and 2.9 V for interval time of 1 us , 1 ms and 1 s , respectively and $(\diamond) 2$-stage square-triangular pulse with square pulse amplitude being modulated in the same way as in ( $\square$ ). (c) Random bit stream generated with interval time $=1 \mathrm{~s}$ using the 2 -stage square-triangular waveform with modulated square pulse amplitude ( $\diamond$ ).

Therefore, external conditions that affects the localization process should also change the relaxation rate. The sandwiched waveform used in Fig. 3c (with a fixed pulse interval of $1 \mu \mathrm{~s}$ ) is repeated for 100 times under different temperatures ( $\mathrm{T}_{\text {emp }}$, relax) (Fig. 4a) or with varying relaxation bias $\mathrm{V}_{\text {relax }}$ (Fig. 4b). It shows that measuring at higher temperatures or applying a bias voltage during the interval could both accelerate the relaxation.

Although $\mathrm{V}_{\mathrm{th}}$ recovery can be accelerated at higher temperature and/or with a relaxation bias, those methods are not convenient to implement in practical TRNG circuits, either requiring heating components or increasing power consumption. A practical solution might be obtained from the operational perspective, by optimizing the bit-generation waveform and modulating the pulse conditions: (i) the higher $\mathrm{V}_{\text {th }}$ caused by the relaxation at longer interval times could be compensated by using a higher square pulse amplitude, inspired by the dependence of $\mathrm{V}_{\text {th }}$ on the bias condition as we have reported in [23], [25]; and (ii) in order to further remove the cumulative effect of failure-to-switch-on, a 2 -stage "squaretriangular" waveform, in which the current is measured at the end of the square pulse, is used to ensure that OTS is switched on after generating each bit, no matter " 0 " or " 1 ", (Fig. 5a), avoiding the need for complex peripheral circuitry to distinguish whether the OTS has been switched on or not during the square stage.

To verify this method, a sequence of 1000 square-triangular waveform pulses with higher square pulse for longer interval time, i.e. modulated square pulse, is applied to this OTS device. The square pulse amplitudes are $2.7 \mathrm{~V}, 2.8 \mathrm{~V}$ and 2.9 V for interval time of $1 \mu \mathrm{~s}, 1 \mathrm{~ms}$ and 1 s , respectively. The triangular pulse starts from the end of the square pulse, reaches its peak of 5 V , and then falls to $0 \mathrm{~V}\left(\mathrm{t}_{\text {fall }}=1 \mu \mathrm{~s}\right)$. Fig. 5 b

TABLE I
NIST Test Result

| Test | p-value | Proportion | Result |
| :---: | :---: | :---: | :---: |
| Frequency | 0.739918 | $10 / 10$ | Pass |
| Block Frequency | 0.534146 | $10 / 10$ | Pass |
| Runs | 0.739918 | $10 / 10$ | Pass |
| Longest Run | 0.534146 | $10 / 10$ | Pass |
| FFT | 0.350485 | $10 / 10$ | Pass |
| Cumulative Sums | $0.350485^{*}$ | $10 / 10$ | Pass |
| Linear Complexity | 0.534146 | $10 / 10$ | Pass |
| Aproximate entropy | 0.739918 | $10 / 10$ | Pass |
| Non- Overlapping | $0.066882^{*}$ | $10 / 10$ | Pass |
| Template | $0.122325^{*}$ | $10 / 10$ | Pass |
| Overlapping Template | $0.066882^{*}$ | $10 / 10$ | Pass |
| Serial | 0.350485 | $1 / 1$ | Pass |
| Rank | The sequence is divided into 10 streams, except the Rank test |  |  |
| *Smallest of the multiple p-values |  |  |  |

demonstrates that, by using the modulated square pulses, the probability shift caused by longer interval is reduced significantly and using the 2 -stage square-triangular pulse with modulated square pulses could effectively keep the switching probability at around $50 \%$.

The randomness of the bits produced was evaluated by the National Institute of Standards and Technology (NIST) Test Suite, a statistical package to evaluate the randomness of binary sequences. Table I summarized the test result for the 10,000 random bits generated by the TRNG, using the 2 -stage square-triangular pulse with modulated square pulse amplitude, as reported in Fig. 5b. The interval time is 1 s . Each test calculates a p -value, and $\mathrm{p}>0.001$ and success proportion of $9 / 10$ are considered good performance. Note that some of the tests consist of several individual tests and we report the smallest p -value out of them. The generated random sequences have passed 12 NIST tests, supporting an excellent randomness performance of the proposed TRNG.

It should be noted that applying a higher pulse based on longer interval time setup, or sending a 2 -stage waveform, is much simpler than designing a complex peripheral circuitry with counters and registers which increase the layout overhead and also cause major reduction in the throughput of random bit generation, as the data post-processing can be time consuming. Our result shows that although the $\mathrm{V}_{\text {th }}$ relaxation is an unavoidable nature of chalcogenide materials, its impact could be conveniently controlled using waveform optimization and condition modulation.

## IV. Conclusion

In this work, we statistically look into the mechanism of switching probability drift in OTS-based TRNG, and have revealed that such switching probability drift can be caused by the relaxation phenomenon in chalcogenide materials after switching-on. Experimental evidence has revealed that relaxation can be accelerated with external bias/temperature, but optimizing the bit-generation waveform and modulating the pulse conditions might be a more practical solution. This work paves the way towards the practical design and operation of OTS-based TRNGs.

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